

Robust learning-based MPC for nonlinear constrained systems

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Abstract

This paper presents a robust learning-based predictive control strategy for nonlinear systems subject to both input and output constraints, under the assumption that the model function is not known a priori and only input-output data are available. The proposed controller is obtained using a nonparametric machine learning technique to estimate a prediction model. Based on this prediction model, a novel stabilizing robust predictive controller without terminal constraint is proposed. The design procedure is purely based on data and avoids the estimation of any robust invariant set, which is in general a hard task. The resulting controller has been validated in a simulated case study.

Key words: Predictive control; Learning control; Robust stability; Nonlinear systems; Lyapunov stability

1 Introduction

Model-based control design, and particularly model predictive control (MPC), relies on the availability of an accurate description of the plant. When a model of the plant dynamics is unavailable a priori, system identification methods can be employed to devise such models automatically from observational data. The objective of this paper is to design a predictive controller based on such a learning method. In this setting, the learning method should be flexible enough to learn rich classes of dynamical systems, while at the same time, it should offer bounds on its predictive performance. The latter is important in the predictive control setting if one wishes to give guarantees on the performance and feasibility of the data-based controller, e.g. for nonlinear [1] or cyber-physical systems [2].

Learning and data-driven predictive controllers have recently gained the attention of the control community [3]. An approach to learning-based MPC that is independent of the concrete learning paradigm was proposed in [4]. A broad scope opens up when considering the learning methodology: some research consider direct weight optimization methods [5,6], others Gaussian processes [7,8], or random forests [9], among many others. Several previous works on this topic have used nonlinear set membership (NSM) methods [10] for learning, like [11,12]. In previous works,

the authors proposed to use kinky inference methods [13] in a model predictive control setting. Kinky inference (KI) methods encompass Lipschitz interpolation [11, 14] and NSM methods [12], and they have several properties that are interesting for MPC. A modified version of this method, tailored to model predictive control, was proposed and used to design a MPC with guaranteed closed-loop properties for systems subject to input constraints [15]. The main limitation of this controller is that it cannot guarantee robust satisfaction of output constraints.

In this paper, the main contribution is a new robust predictive controller for systems that also have output constraints. A design method that takes into account the prediction error bounds in an explicit way to tighten the problem constraints and that guarantees closed-loop constraint satisfaction and input-to-state stability (ISS) [16] is provided. One of the main characteristics of this design is that it is not based on a terminal region constraint. In general, this terminal constraint is based on robust invariant sets which, for the class of systems considered (that is, unknown systems, possibly nonlinear, for which a priori only input/output data is available), are difficult to obtain.

Another contribution of this work is the procedure to define the tightened constraints of the MPC optimization problem. They are specifically tailored to the inference model used, in order to obtain the least conservative possible bounds. In contrast to the preliminary version of this controller, pre-

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sented in [17], an output-feedback formulation is considered, and the stability analysis is addressed.

Notation: For two column vectors v, w , (v, w) implies $[v^T, w^T]^T$. Given two sets A, B , $A \oplus B$ is the Minkowski sum and $A \ominus B$ the Pontryagin difference. The set of the integers in the interval $[a, b]$ is denoted \mathbb{I}_a^b . A function $\alpha : \mathbb{R}_{\geq 0} \rightarrow \mathbb{R}_{\geq 0}$ is a \mathcal{K} -function if it is continuous, strictly increasing and $\alpha(0) = 0$. Given a vector $v \in \mathbb{R}^{n_y}$, the ball $\mathcal{B}(v) \subset \mathbb{R}^{n_y}$ is defined as $\mathcal{B}(v) = \{y : |y_s| \leq v_s, s \in \mathbb{I}_1^{n_y}\}$; and $|v|$ denotes the vector whose components are the absolute value of the components of v . I_n is an identity matrix of size n . Given a compact set Ω , $\|\Omega\|_q = \max_{x \in \Omega} \|x\|_q$, for some norm $\|\cdot\|_q$.

2 Problem setting

In this paper it is assumed that the system to be controlled is a sampled continuous-time system described by an a-priori unknown discrete-time model, where $y(k) \in \mathbb{R}^{n_y}$ is its measured output and $u(k) \in \mathbb{R}^{n_u}$ is the control input. Both inputs and outputs are subject to hard constraints

$$u(k) \in \mathcal{U}, \quad y(k) \in \mathcal{Y}, \quad (1)$$

where both \mathcal{U} and \mathcal{Y} are compact sets. It is assumed, without loss of generality, that the origin is the equilibrium point of the system where the plant must be stabilized.

It is assumed that the only information available from the plant is historical data, containing a certain set of measured inputs and outputs trajectories, \mathcal{D} . The objective of the paper is to design an output-feedback control law

$$u(k) = \kappa(y(k); \mathcal{D}), \quad (2)$$

such that from the data set \mathcal{D} and the current output measurement $y(k)$, the control action is computed. It is desired to devise the control law such that the closed-loop system is asymptotically stable and that the constraints are satisfied for all time steps $k \in \mathbb{N}$.

Since a model of the dynamics is not available a priori, it is assumed that the measured output can be used to describe the model of the system with the following nonlinear autoregressive exogenous (NARX) model of the plant [18, 19]:

$$y(k+1) = f(x(k), u(k)) + e(k), \quad (3)$$

where $x(k) = (y(k), \dots, y(k - n_a), u(k - 1), \dots, u(k - n_b)) \in \mathcal{X} := \mathcal{Y}^{(n_a+1)} \times \mathcal{U}^{n_b} \subseteq \mathbb{R}^{n_x}$ with $n_x = (n_a + 1)n_y + n_b n_u$, for some memory horizon lengths $n_a, n_b \in \mathbb{N}$. The residual $e(k)$ models process noise and it is assumed to be confined to a compact set $\mathcal{E} \subset \mathbb{R}^{n_y}$. For notational convenience, the inputs of f are aggregated into a joint vector $w := (x, u) \in \mathcal{W}$, which is referred to as *regressor*.

Remark 1 The horizons n_a and n_b represent the model order. In [19], the conditions under which they could be taken as $n_a = n_b = 2n$ were given, where n is the order of the system. If n is unknown, one would have to use the best guess, or cross-validation methods to estimate n_a and n_b .

Assumption 1 (Hölder continuity) Each output component of the function $f(\cdot)$, referred to as ground truth function, is Hölder continuous. That is, there exist some constants $L_{f,i} > 0$ and $p_{f,i} \in (0, 1]$ ($i \in \mathbb{I}_1^{n_y}$) such that $\forall w_1, w_2 \in \mathcal{W}$

$$|f_i(w_1) - f_i(w_2)| \leq L_{f,i} \|w_1 - w_2\|_{\mathcal{W}}^{p_{f,i}}, \quad i \in \mathbb{I}_1^{n_y}, \quad (4)$$

where $\|\cdot\|_{\mathcal{W}}$ stands for a specific norm defined for the regressors, and the sub-index i denotes the i -th component of the vector. For each $i \in \mathbb{I}_1^{n_y}$, any constant $L_{f,i}$ that satisfies this condition is called a Hölder constant, while the lowest of them is called the best Hölder constant, $L_{f,i}^*$.

Remark 2 Assumption 1 can be relaxed to general continuity, provided that both \mathcal{Y} and \mathcal{U} are compact sets.

2.1 The learning method

In this section the machine learning method used to estimate f is presented, sometimes called kinky inference [13]. Using the available experimental data, a data set \mathcal{D} of $N_{\mathcal{D}}$ regressor/outputs is collected; that is

$$\mathcal{D} := \{(y(j), w(j)) \mid j = 1, \dots, N_{\mathcal{D}}\}. \quad (5)$$

The structure of \mathcal{D} depends on the value of n_a and n_b . To predict an unseen query point w , KI makes use of the data base \mathcal{D} , and provides an estimation of the Hölder parameters, denoted L and p , which are vectors of dimension n_y .¹

Definition 1 (Kinky inference rule) The i -th output component function of the KI predictor, for $i = 1, \dots, n_y$, shall be defined by

$$\hat{f}_i(w; L_i, p_i, \mathcal{D}) := \frac{1}{2} \min_{j=1, \dots, N_{\mathcal{D}}} y_{j,i} + L_i \|w - w_j\|_{\mathcal{W}}^{p_i} + \frac{1}{2} \max_{j=1, \dots, N_{\mathcal{D}}} y_{j,i} - L_i \|w - w_j\|_{\mathcal{W}}^{p_i}. \quad (6)$$

Given the hyperparameters L and p and the data set \mathcal{D} , the predictor \hat{f} of the ground-truth f is constructed, yielding the output prediction:²

$$\hat{y}(k+1) = \hat{f}(x(k), u(k)). \quad (7)$$

¹ In order to reduce the estimation error, different Hölder constants L_i are used to estimate the i -th entry of the output vector. With a slight abuse of notation L and p will denote a vector in \mathbb{R}^{n_y} .

² For the sake of conciseness, the dependence of \hat{f} with L, p and \mathcal{D} may be omitted in the rest of the paper.

This predictor \hat{f} is Hölder continuous, and as proven in [20, Lemma 5], the Hölder parameters of f are also Hölder parameters of \hat{f} . That is, for a given exponent p , the Lipschitz constant L^* is also a Lipschitz constant of \hat{f} . However, the estimated L and p will suffice to derive the stability properties of the proposed controller, as it will be explained later.

Remark 3 While in several works a priori knowledge of the correct parameters L_f and p_f is assumed [10, 21], other works provide methods of calculating these parameters from the available data [12, 20]. In this paper, the so-called LACKI method [20] is applied. It obtains the Lipschitz constant L as the minimum one that is consistent with the data. The learning feature of the proposed predictor is also demonstrated in [20]. Note that, w.l.o.g., this paper extends the LACKI method to each output component function in isolation.

In this paper, the paradigm of predictive control will be employed to derive the data-based control law (2). MPC requires repeated optimization of the predicted control inputs subject to constraints. Therefore, in order to give guarantees on the controller's closed-loop performance, recursive feasibility and constraint satisfaction must be ensured. That is, it is necessary to ensure that all constraints remain satisfiable during runtime, or equivalently, to guarantee that the controlled system will not leave the feasibility region. However, since the controller will not be based on the ground-truth dynamics f , but on the learned model \hat{f} inferred from a sample of the ground-truth, recursive feasibility can only be guaranteed if a bound on the discrepancy between f and \hat{f} is known a priori and taken into account by the controller.

The estimation method ensures that if the model function is Hölder and the noise is bounded, then the estimation error is bounded [20], which is required to design a deterministic robust controller to regulate the plant. Any worst-case guarantee inevitably requires a priori knowledge. Hence, in the following hypothesis, it is assumed that this bound is available for the design of the controller.

Assumption 2 It is assumed that for L , p , and \mathcal{D} , a bound on the error between the estimated output and the real output is known, denoted $\mu \in \mathbb{R}^{n_y}$, such that

$$|\hat{f}_i(x, u) - f_i(x, u) - e_i| \leq \mu_i, \quad (8)$$

for all $i \in \mathbb{I}_1^{n_y}$, $e \in \mathcal{E}$, $x \in \mathcal{Y}^{n_a+1} \times \mathcal{U}^{n_b}$, and $u \in \mathcal{U}$.

Remark 4 From a practical point of view the problem of how to calculate the error bound must be addressed. Kinky inference methods enjoy the property of providing a deterministic error bound if the Lipschitz constant and an upper bound of the noise are known [13]. Moreover, if a bound on the second derivative is known, it is also possible to derive an estimation error bound. If these parameters are not known a priori, which is usual in practice, then they must

be estimated from experimental data. Consequently, the validity of the results presented in this paper is conditioned to the validity of the estimated error bound. This is the reason why this is considered as a standing assumption.

Remark 5 The KI prediction method has recently been improved in [15] decreasing the computational cost and smoothing the prediction, in order to enhance the optimization that will be carried out by the controller.

3 Stabilizing data-based NMPC

In this section, a model predictive controller is derived based on a prediction model learned from data of the plant. Since the prediction model is not accurate, the effect of the estimation error on the predictions must be analysed to be taken into account in the design of the controller. For this analysis, it is convenient to define the NARX model of the plant in a state-space form as follows:

$$x(k+1) = F(x(k), u(k)) + \xi(k) \quad (9a)$$

$$y(k) = Mx(k), \quad (9b)$$

where

$$F(x(k), u(k)) = \left(f(x(k), u(k)), y(k), \dots, y(k-n_a+1), u(k), \dots, u(k-n_b+1) \right), \quad (10)$$

$$M = [I_{n_y}, 0, \dots, 0], \text{ and } \xi(k) = (e(k), 0, \dots, 0).$$

Let $\hat{y}(j|k)$ denote the output that, at time k , is predicted to be observed at time $k+j$, for a given candidate control sequence $u(k+j), j \in \mathbb{I}_0^{N-1}$. Then the predicted state is given by

$$\hat{x}(j|k) = (\hat{y}(j|k), \dots, \hat{y}(1|k), y(k), \dots, y(k+j-n_a), u(k+j-1), \dots, u(k+j-n_b)),$$

where

$$\hat{x}(j+1|k) = \hat{F}(\hat{x}(j|k), u(k+j)) \quad (11)$$

and

$$\begin{aligned} \hat{F}(\hat{x}(j|k), u(k+j)) = & (\hat{f}(\hat{x}(j|k), u(k+j)), \\ & \hat{y}(j|k), \dots, y(k), \dots, \\ & y(k+j-n_a+1), \dots, \\ & u(k+j), \dots, u(k+j-n_b+1)). \end{aligned}$$

The proposed robust MPC is based on nominal predictions and tightened constraints. To guarantee robustness, a bound on the propagation of the prediction error is calculated from the following lemma:

Lemma 1 Assume that at sampling time k , the state of the plant is $x(k)$ and a sequence of future control inputs $u(k+j)$

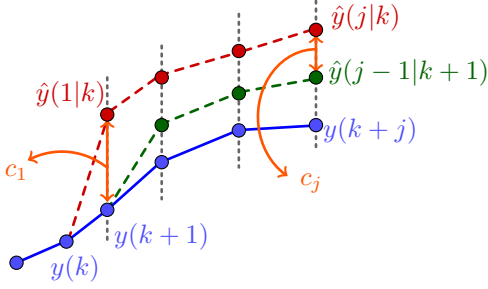


Fig. 1. Propagation of the prediction error

for $j \in \mathbb{I}_0^{N-1}$ is given. Let $\hat{x}(j|k)$ and $\hat{y}(j|k)$ be the predicted states and outputs, respectively, derived from (11) for the given sequence of future control inputs and the current state $x(k)$, i.e. $\hat{x}(0|k) = x(k)$.

Assume that at sampling time $k+1$, the current output $y(k+1)$ is measured, and hence the current state $x(k+1)$ is known. Based on these new measurements, an updated sequence of states and outputs $\hat{x}(j|k+1)$ and $\hat{y}(j|k+1)$ are predicted based on (11) with $\hat{x}(0|k+1) = x(k+1)$ and the remaining sequence of the given future control inputs.

Let $c_1 \in \mathbb{R}^{n_y}$ be a vector such that

$$|y(k+1) - \hat{y}(1|k)| \leq c_1. \quad (12)$$

Then, the mismatch between the predictions satisfies³

$$|\hat{y}(j-1|k+1) - \hat{y}(j|k)| \leq c_j, \quad j \in \mathbb{I}_1^N, \quad (13a)$$

$$\|\hat{x}(j-1|k+1) - \hat{x}(j|k)\|_{\mathcal{X}} \leq r_j, \quad j \in \mathbb{I}_1^N, \quad (13b)$$

where $c_j \in \mathbb{R}^{n_y}$ and $r_j \in \mathbb{R}$ are obtained from the recursion

$$c_{j+1,i} = L_i r_j^{p_i}, \quad (14)$$

and $r_j = \|\Xi_j\|_{\mathcal{X}}$, $j \in \mathbb{I}_1^{N-1}$, $i \in \mathbb{I}_1^{n_y}$, where

$$\Xi_j = \mathcal{B}(c_j) \times \cdots \times \mathcal{B}(c_{\sigma(j)}) \times \underbrace{\{0\} \times \cdots \times \{0\}}_{n_b+1-\sigma(j-1) \text{ times}} \subseteq \mathbb{R}^{n_x},$$

with $\sigma(j) = \max(1, j - n_a)$.

PROOF. Provided that $\hat{y}(j-1|k+1) = \hat{f}(x(j-2|k+1), u(k+j-1))$ and $\hat{y}(j|k) = \hat{f}(\hat{x}(j-1|k), u(k+j-1))$, it can be derived that $\forall i \in \mathbb{I}_1^{n_y}$

$$|\hat{y}_i(j-1|k+1) - \hat{y}_i(j|k)| \leq L_i \|\hat{x}(j-2|k+1) - \hat{x}(j-1|k)\|_{\mathcal{X}}^{p_i}.$$

³ $\|\cdot\|_{\mathcal{X}}$ is a norm for the state-space such that $\|x\|_{\mathcal{X}} = \|(x, 0)\|_{\mathcal{W}}$

Given that

$$\begin{aligned} & \hat{x}(j-2|k+1) - \hat{x}(j-1|k) = \\ & [\hat{y}(j-2|k+1) - \hat{y}(j-1|k), \hat{y}(j-3|k+1) - \hat{y}(j-2|k), \\ & \dots, \hat{y}(\sigma(j-1)-1|k+1) - \hat{y}(\sigma(j-1)|k), 0, \dots, 0]. \end{aligned}$$

then $\hat{x}(j-2|k+1) - \hat{x}(j-1|k) \in \Xi_{j-1}$. Assuming that c_{j-1} is known,

$$\|\hat{x}(j-2|k+1) - \hat{x}(j-1|k)\|_{\mathcal{X}} \leq \|\Xi_{j-1}\|_{\mathcal{X}} = r_{j-1},$$

which implies the stated result. \blacksquare

Remark 6 If the infinity norm is chosen as the norm of the input space then

$$r_j = \max_{s \in \mathbb{I}_{\sigma(j)}^j} \|c_s\|_{\infty}.$$

Based on the derived bounds on the prediction error, the problem of robust constraint satisfaction is addressed by means of a set of tightened constraints on the outputs [22], computed offline for the maximum possible prediction error, i.e. taking $c_1 = \mu$. These sets are defined as follows:

$$\mathcal{Y}_j = \mathcal{Y} \ominus \mathcal{B}(d_j), \quad (15)$$

where

$$d_j = \sum_{s=1}^j c_s. \quad (16)$$

These constraints sets will be used to prove recursive feasibility of the controller, following standard procedures.

Lemma 2 The sets \mathcal{Y}_j are such that for all $y \in \mathcal{Y}_j$ and for all $\Delta y \in \mathcal{B}(c_j)$, $y + \Delta y \in \mathcal{Y}_{j-1}$.

PROOF. Since for $j \geq 1$, $d_j = d_{j-1} + c_j$, it follows that

$$\mathcal{B}(d_j) = \mathcal{B}(d_{j-1}) \oplus \mathcal{B}(c_j).$$

By definition,

$$y + \Delta y \in \mathcal{Y}_j \oplus \mathcal{B}(c_j) = \mathcal{Y} \ominus \mathcal{B}(d_j) \oplus \mathcal{B}(c_j),$$

and hence $\mathcal{Y}_j = \mathcal{Y} \ominus \mathcal{B}(d_j) = \mathcal{Y} \ominus \mathcal{B}(d_{j-1}) \ominus \mathcal{B}(c_j)$, so

$$\begin{aligned} y + \Delta y & \in \mathcal{Y}_j \oplus \mathcal{B}(c_j) \\ & = \mathcal{Y} \ominus \mathcal{B}(d_{j-1}) \ominus \mathcal{B}(c_j) \oplus \mathcal{B}(c_j) \\ & \subseteq \mathcal{Y} \ominus \mathcal{B}(d_{j-1}) = \mathcal{Y}_{j-1}. \quad \blacksquare \end{aligned}$$

In order to ensure that the proposed controller is feasible, the tightened set of constraints must be non-empty along the prediction horizon, as stated in the following assumption:

Assumption 3 The prediction horizon N and the estimation error bound μ are such that the set \mathcal{Y}_N is non-empty.

Based on the previous definitions, the optimization problem $P_N(x(k); \mathcal{D})$ of the proposed predictive controller is:

$$\begin{aligned} \min_{\mathbf{u}} V_N(x(k), \mathbf{u}) \\ &= \sum_{i=0}^{N-1} \ell(\hat{x}(i|k), u(i)) + \lambda V_f(\hat{x}(N|k)) \quad (17a) \\ \text{s.t. } \hat{x}(0|k) &= x(k) \quad (17b) \\ \hat{x}(j+1|k) &= \hat{F}(\hat{x}(j|k), u(j)), \quad j \in \mathbb{I}_0^{N-1} \quad (17c) \\ \hat{y}(j|k) &= M\hat{x}(j|k), \quad j \in \mathbb{I}_0^{N-1} \quad (17d) \\ u(j) &\in \mathcal{U}, \quad j \in \mathbb{I}_0^{N-1} \quad (17e) \\ \hat{y}(j|k) &\in \mathcal{Y}, \quad j \in \mathbb{I}_0^{N-1}, \quad (17f) \end{aligned}$$

where $\lambda \geq 1$ is a weighting parameter. Note that this problem is non-linear, non-convex and non-differentiable. Its ingredients are required to meet the following assumption, which is similar to the standard MPC ones [22]:

Assumption 4

- (1) The stage cost function $\ell(\cdot, \cdot)$ is a Hölder continuous positive definite function such that $\ell(\cdot, \cdot) \geq \alpha_y(\|\cdot\|_{\mathcal{X}})$ for a certain \mathcal{K} -function α_y , and its Hölder parameters are L_x and p_x .
- (2) There exists a control law $u = \kappa_f(x)$, a function V_f and a level set $\Omega_\gamma = \{x : V_f(x) \leq \gamma\} \subseteq \mathbb{R}^{n_x}$ for some $\gamma > 0$ such that for all $x \in \Omega_\gamma$ the following conditions hold:
 - (a) V_f is a Hölder continuous positive definite function, with Hölder constants L_{V_f}, p_{V_f} , such that
$$\begin{aligned} \alpha_f(\|x\|_{\mathcal{X}}) &\leq V_f(x) \leq \beta_f(\|x\|_{\mathcal{X}}), \\ V_f(\hat{F}(x, \kappa_f(x))) - V_f(x) &\leq -\ell(x, \kappa_f(x)). \end{aligned}$$
 - (b) $\kappa_f(x) \in \mathcal{U}$, $Mx \in \mathcal{Y}_N$.

The controller is derived from the receding horizon solution of (17). It follows a standard robust approach in which the cost of the nominal predictions is minimized, while taking into account a tightened set of constraints to guarantee recursive feasibility. The main difference with off-the-shelf robust ISS formulations for nonlinear systems [16] is that in these, either there are no constraints on the states in the optimization problem, or a terminal constraint, based on a certain robust positive invariant set, is added. In this controller, although a terminal cost (based on a local controller for the nominal model) is taken into account in the cost function, no terminal constraint is included. Thus, its design is notably simplified since the calculation of a robust invariant set is avoided, which was a hard task, as shown in [23]. In this case, the calculation could have been even more difficult provided the lack of an explicit expression of the model of the system.

Furthermore, an additional tuning parameter λ is added, modifying the weight of the terminal cost in the objective

function. It is proven that this controller guarantees that the closed-loop systems is ISS in an explicitly defined region of the state space, which is enlarged by this weight.

Define the function

$$\nu(c_1) = \sum_{j=1}^N L_x r_j^{p_x} + \lambda L_{V_f} r_{N+1}^{p_{V_f}},$$

where r_j is defined in Lemma 1 for c_1 , and L_x, L_{V_f} in Ass. 4.

Assumption 5

The bound μ is such that the set $\Upsilon = \{x : \ell(x, 0) \leq \nu(\mu)\}$ is contained in Ω_γ . The positive constants λ and ϕ are such that $\lambda \geq 1$ and $\ell(x, 0) > \phi$ for all $x \notin \Omega_\gamma$.

Remark 7 In a general setting, a condition to check if the level set Υ is contained in Ω_γ could be derived using the supply \mathcal{K}_∞ -functions that bound the cost functions given in Assumption 4. In this case the condition would be:

$$\nu(\mu) \leq \alpha_y(\beta_f^{-1}(\gamma)). \quad (18)$$

Another method could be using probabilistic validation by means of randomized algorithms [24].

Lemma 3 Under Assumption 5, $\phi \geq \nu(\mu)$.

PROOF. Since $\Upsilon \subseteq \Omega_\gamma$, the constant ϕ can be taken as $\phi \geq \min_{x \in \mathcal{X} \setminus \Omega_\gamma} \ell(x, 0) \geq \min_{x \in \mathcal{X} \setminus \Upsilon} \ell(x, 0) \geq \nu(\mu)$. ■

Let Γ define the following level set of the optimal cost function

$$\Gamma = \{x : V_N^*(x) \leq N\phi + \lambda\gamma\}.$$

It is next proven that this set defines the region in which ISS is guaranteed. Notice that this set is compact and non-empty.

Theorem 1 (ISS stability) Suppose that Assumptions 2, 3, 4 and 5 hold for the optimization problem $P_N(\cdot)$. Let $\kappa_N(x)$ be the control law derived from the solution of $P_N(x; \mathcal{D})$ applied using a receding horizon policy. Then, for any $x(0) \in \Gamma$, the system controlled by the control law $u(k) = \kappa_N(x(k))$ is input-to-state stable w.r.t. the estimation error; and the constraints are always satisfied, i.e. $u(k) \in \mathcal{U}$, $y(k) \in \mathcal{Y}$ and $x(k) \in \Gamma, \forall k$.

PROOF. Assume that $x(k) \in \Gamma$. Then, it can be shown that $x^*(N|k) \in \Omega_\gamma$ [25]. Define the shifted sequence as $\bar{u}(k+1)$ such that $\bar{u}(j|k+1) = u^*(j+1|k)$ for $j \in \mathbb{I}_0^{N-2}$ and $\bar{u}(N-1|k+1) = \kappa_f(x^*(N|k))$.

Recursive feasibility: Assuming that $x(k) \in \Gamma$, it will be proven that $x(k+1) \in \Gamma$. Since Γ is a subset of the feasibility region of the optimization problem $P_N(x; \mathcal{D})$, the system is recursively feasible.

Firstly, it will be shown that the solution $\bar{u}(k+1)$ is a feasible solution for $x(k+1)$. Given that $x^*(N|k) \in \Omega_\gamma$, from the

feasibility of $\mathbf{u}^*(k)$ and assumption 4, it is immediate to state that $\bar{u}(j|k+1) \in \mathcal{U}$ for all $j \in \mathbb{I}_0^{N-1}$.

From Lemma 1, $\hat{y}(j|k+1) - y^*(j+1|k) \in \mathcal{B}(c_{j+1})$, $\forall j \in \mathbb{I}_0^{N-1}$ and from the feasibility of $\mathbf{u}^*(k)$. Thus, $y^*(s|k) \in \mathcal{Y}_s$ for $s \in \mathbb{I}_0^{N-1}$ and $x^*(N|k) \in \Omega_\lambda$, which implies that $y^*(N|k) = Mx^*(N|k) \in \mathcal{Y}_N$ in virtue of assumption 4.2b. Then, from Lemma 2, for all $j \in \mathbb{I}_0^{N-1}$,

$$\hat{y}(j|k+1) \in \mathcal{Y}_{j+1} \oplus \mathcal{B}(c_{j+1}) \subseteq \mathcal{Y}_j.$$

Therefore the problem $P_N(x(k+1); \mathcal{D})$ is feasible.

Next, it will be proven that $x(k+1) \in \Gamma$. Since $x(k) \in \Gamma$, $V_N^*(x(k)) \leq N\phi + \lambda\gamma$. Following standard arguments in MPC [22] it can be proven that

$$\begin{aligned} V_N(x^*(1|k), \bar{\mathbf{u}}(k+1)) &\leq V_N^*(x(k)) - \ell(x(k), u(k)) \\ &\leq N\phi + \lambda\gamma - \ell(x(k), u(k)). \end{aligned}$$

On the other hand, given $\tilde{x} = \hat{F}(x^*(N|k), \kappa_f(x^*(N|k)))$,

$$\begin{aligned} V_N(x(k+1), \bar{\mathbf{u}}(k+1)) - V_N(x^*(1|k), \bar{\mathbf{u}}(k+1)) &= \\ &\sum_{i=0}^{N-1} \left(\ell(\hat{x}(i|k+1), \bar{u}(i|k+1)) \right. \\ &\quad \left. - \ell(x^*(i+1|k), \bar{u}(i|k+1)) \right) \\ &\quad + \lambda \left(V_f(\hat{x}(N|k+1)) - V_f(\tilde{x}) \right). \end{aligned}$$

According to Lemma 1, it can be derived that for $s \in \mathbb{I}_1^N$, $\|\hat{x}(s-1|k+1) - x(s|k)\|_{\mathcal{X}} \leq r_s$, with r_s obtained for a given c_1 . Then, for $j \in \mathbb{I}_1^N$ and given c_1 satisfying (12),

$$\begin{aligned} \ell(\hat{x}(j-1|k+1), \bar{u}(j-1|k+1)) \\ - \ell(x^*(j|k), \bar{u}(j-1|k+1)) &\leq L_x r_j^{p_x} \end{aligned}$$

$$\text{and } V_f(\hat{x}(N|k+1)) - V_f(\tilde{x}) \leq L_{V_f} r_{N+1}^{p_{V_f}}.$$

Therefore,

$$V_N(x(k+1), \bar{\mathbf{u}}(k+1)) - V_N(x^*(1|k), \bar{\mathbf{u}}(k+1)) \leq \nu(c_1). \quad (19)$$

To prove robust invariance the worst possible case has to be considered, for which $c_1 = \mu$ is taken. Hence, it has been proven that

$$V_N(x(k+1), \bar{\mathbf{u}}(k+1)) \leq \nu(\mu) + N\phi + \lambda\gamma - \ell(x(k), u(k)).$$

Consider the case where $x(k) \in \Gamma \setminus \Upsilon$. Then $\ell(x(k), u(k)) > \nu(\mu)$. Hence, $V_N(x(k+1), \bar{\mathbf{u}}(k+1)) \leq \nu(\mu) + N\phi + \lambda\gamma - \ell(x(k), u(k)) \leq N\phi + \lambda\gamma$.

Given that $V_N^*(x(k+1)) \leq V_N(x(k+1), \bar{\mathbf{u}}(k+1))$ then $x(k+1) \in \Gamma$.

Consider now the case that $x(k) \in \Upsilon$. Since $\Upsilon \subseteq \Omega_\gamma$, $x(k) \in \Omega_\gamma$. With standard arguments in MPC [22], it can be shown that $V_N^*(x(k)) \leq \lambda V_f(x(k)) \leq \lambda\gamma$. Hence,

$$\begin{aligned} V_N(x(k+1), \bar{\mathbf{u}}(k+1)) &\leq \nu(\mu) + V_N^*(x(k)) \\ &\quad - \ell(x(k), u(k)) \\ &\leq \nu(\mu) + \lambda\gamma - \ell(x(k), u(k)), \end{aligned}$$

since $\nu(\mu) \leq \phi$, $V_N(x(k+1), \bar{\mathbf{u}}(k+1)) \leq N\phi + \lambda\gamma$. Thus, $x(k+1) \in \Gamma$.

Input-to-state stability: Equation (19) can be rewritten as follows, taking $c_1 = e_p(k+1) := |y(k+1) - \hat{y}(1|k)|$:

$$V_N(x(k+1), \bar{\mathbf{u}}(k+1)) - V_N(x^*(1|k), \bar{\mathbf{u}}(k+1)) \leq \nu(e_p(k+1)). \quad (20)$$

Then, following the previous steps, it can be derived that

$$\begin{aligned} V_N^*(x+1) &\leq V_N(x(k+1), \bar{\mathbf{u}}(k+1)) \\ &\leq \nu(e_p(k+1)) + V_N^*(x(k)) - \ell(x(k), u(k)). \end{aligned} \quad (21)$$

Thus, $V_N^*(x)$ is an ISS Lyapunov function [16]. \blacksquare

Remark 8 (Suboptimal case) *The stability analysis can be extended to the case in which the optimal solution of the control problem is not found. Given an initial feasible solution of the control problem, if the optimiser is able to improve the cost (even for a suboptimal solution of the problem), then the controller is able to maintain robust stability while satisfying the constraints [22].*

Remark 9 (Violation of Assumption 2) *If the bound on the prediction error is estimated from data (e.g. via cross-validation), then the error could take a value larger than μ for a certain period of time. In this case, the ISS property (21) still holds as long as $x(k) \in \Gamma$ for that period of time. Notice that the ISS condition is derived from the smoothness of the optimal cost function, which is an inherent property of the proposed optimal control problem.*

Remark 10 (Stability margin) *Most of the robust controllers for constrained systems exhibit an upper bound on the estimation error to be solvable [16]. This is the so-called stability margin. One of the main drawbacks of robust predictive controllers is that this margin is typically quite conservative, due to the open-loop nature of the predictions. As a robust controller, our approach inherits this drawback.*

4 Case study

In this section, the proposed controller is applied to a continuously-stirred tank reactor [15], in which a reaction $A \rightarrow B$ takes place. The system's state is defined by the concentration of the reactant, C_A (mol/l), the temperature in the tank, T (K), and the temperature of the coolant, T_c (K). The state varies with respect to the control input,

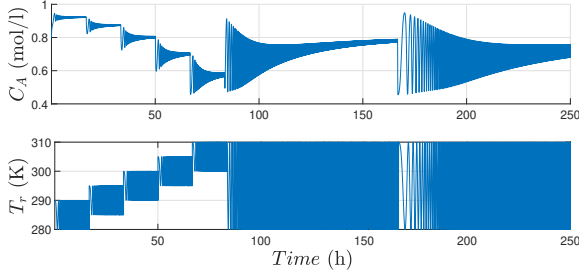


Fig. 2. Simulation obtained applying a sequence of chirp signals

which is the reference temperature of the coolant, T_r (K), according to the following set of ordinary differential equations (ODEs), which will be used to simulate the system but are assumed to be unknown:

$$\frac{dC_A(t)}{dt} = \frac{q_0}{V} \cdot (C_{Af} - C_A(t)) - k_0 \cdot e\left(-\frac{E}{R \cdot T(t)}\right) \cdot C_A(t), \quad (22a)$$

$$\frac{dT(t)}{dt} = \frac{q_0}{V} \cdot (T_f - T(t)) + \frac{(-\Delta H_r) \cdot k_0}{\rho \cdot C_p} \cdot e\left(-\frac{E}{R \cdot T(t)}\right) \cdot C_A(t) + \frac{U \cdot A}{V \cdot \rho \cdot C_p} \cdot (T_c(t) - T(t)), \quad (22b)$$

$$\frac{dT_c(t)}{dt} = \frac{T_r(t) - T_c(t)}{\tau}. \quad (22c)$$

The parameters of the system can be found in [26], with $\tau = 1.5$ min. The sampling time is 30 s. The output is C_A , and its sensor has a 2.5% error margin, generated randomly using a uniform distribution. The constraints are given by $0.38 \leq C_A \leq 0.954$ mol/l and $280 \leq T_r \leq 310$ K. The reference equilibrium point is given by $C_A^{\text{ref}} = 0.62$ mol/l and $T_r^{\text{ref}} = 304.5$ K.

In order to identify the system, several data sets are generated. The training data set is obtained using the following input sequence of chirp signals: five chirp signals of length 1000 min, initial and final frequencies of 1 mHz and 0.15 Hz (respectively), amplitude 5 K, and centres starting from 285 K with 5 K interval; followed by two chirp signals of length 5000 min, centered in 295 K, 15 K of amplitude, and initial and final frequencies of 10 and 200 mHz and 1 and 90 mHz, respectively, as represented in Figure 2.

Another data set is obtained to calculate μ via cross-validation, applying a pseudorandom binary input sequence, where T_r switches randomly between 280 and 310 K with switching periods between 12 and 75 min.

Following standard cross-validation procedures [27], these data sets are used to define a predictor for different values of the memory horizons n_a and n_b . Setting the prediction horizon to $N = 4$ and applying LACKI [20] to calculate L while

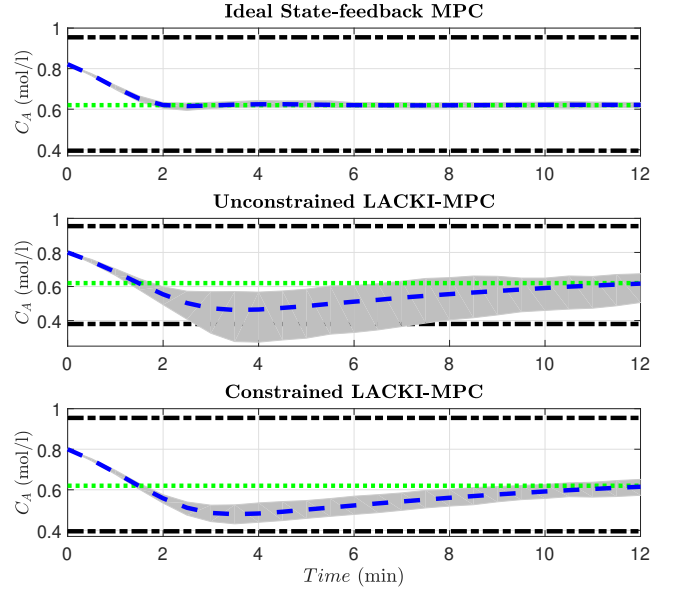


Fig. 3. Closed-loop output for 100 simulations of the ideal MPC (top), the robust learning MPC for systems without output constraints (middle) and the proposed robust constrained learning MPC (bottom). The grey band groups the trajectories, the blue dashed line represents its mean, the green dotted one the reference and the black dash-dotted one the constraints

fixing $p = 1$, the optimal n_a, n_b are obtained minimising d_4 , which is calculated using (16). The minimum is obtained for $n_a = n_b = 2$, for which $L = 1.38$, $\mu = 0.032$ mol/l and $d_4 = 0.22$ mol/l.

Both the stage and the terminal cost of the MPC are defined as follows, with $x^{\text{ref}} = (y^{\text{ref}}, \dots, y^{\text{ref}}, u^{\text{ref}}, \dots, u^{\text{ref}})$:

$$\ell(x, u) = \|x - x^{\text{ref}}\|_Q^2 + \|u - u^{\text{ref}}\|_R^2, \quad (23a)$$

$$V_f(x) = \|x - x^{\text{ref}}\|_P^2. \quad (23b)$$

Q is set to 100, $R = 0.1$, and $\lambda = 10$. Using the model with $n_a = n_b = 2$ the terminal cost is obtained solving a LQR for the linearised model around the reference point. To ensure robust stability (Theorem 1), Assumptions 2, 3, 4 and 5 must hold true. The prediction error was obtained via cross-validation. The value of d_N results in $\mathcal{Y}_N = \{y : 0.60 \leq y \leq 0.73\}$. Following the procedure in section 3 and in [15] results in $\gamma = 16796$, $\nu(\mu) = 53.553$ and $\phi = 9.2189 \times 10^5$, which satisfy all the assumptions.

The proposed controller is applied in 100 closed-loop simulations, subject to random noise. The results are shown in the last row of Figure 3. Note that the output is steered to the reference while the constraints are satisfied. The optimization problem is solved in Matlab on an Intel® Core™ i7-6700HQ CPU @ 2.60 GHz 12GB RAM and each iteration takes less than one second to complete, much shorter than the 30 s required by the sampling time.

In order to compare the proposed controller to other methods, the same setup is simulated with two different MPCs. First, a controller derived from (17), but with the set of ODEs (22) as the state-feedback prediction model, for which μ is the maximum noise, 0.025 mol/l. This aims to resemble the ideal case of perfect knowledge of the plant (result shown in the first row of Fig. 3). As expected, the data-based controller performs slower than the ideal since, unlike the latter, an output-feedback uncertain framework is considered.

Second, the controller proposed in [15] is applied. It is based on KI and guarantees closed-loop stability but does not take into account output constraints, so as shown in the second row of Fig. 3, the minimum C_A limit is violated. To sum up, the controller proposed in this paper is able to robustly satisfy hard constraints in the outputs, learning the model from input-output data with a closed-loop performance similar to the ideal case.

5 Conclusion

A novel learning-based predictive controller capable of ensuring robust stability without terminal constraint was proposed and proven to be robustly stable (in the ISS sense) and recursively feasible under some assumptions. The plant model of this controller is learned from input-output data, using the LACKI approach. Under assumptions on conservatism of estimated maximum bounds, tight bounds on the effect of the multiple step look-ahead uncertainty were derived and ISS stability was proven. Simulation of a continuously-stirred tank reactor illustrated the practical feasibility of the proposed controller.

References

- [1] F. Allgöwer and A. Zheng, *Nonlinear model predictive control*. Birkhäuser, 2012, vol. 26.
- [2] M. Behl, A. Jain, and R. Mangharam, “Data-driven modeling, control and tools for cyber-physical energy systems,” in *Proceedings of the 7th International Conference on Cyber-Physical Systems*. IEEE Press, 2016, p. 35.
- [3] L. Hewing, K. P. Wabersich, M. Menner, and M. N. Zeilinger, “Learning-based model predictive control: Toward safe learning in control,” *Annual Review of Control, Robotics, and Autonomous Systems*, vol. 3, 2019.
- [4] A. Aswani, H. Gonzalez, S. S. Sastry, and C. Tomlin, “Provably safe and robust learning-based model predictive control,” *Automatica*, vol. 49, no. 5, pp. 1216–1226, 2013.
- [5] D. Piga, S. Formentin, and A. Bemporad, “Direct data-driven control of constrained systems,” *IEEE Transactions on Control Systems Technology*, vol. 26, no. 4, pp. 1422–1429, 2017.
- [6] J. R. Salvador, D. R. Ramirez, T. Alamo, and D. Muñoz de la Peña, “Offset free data driven control: application to a process control trainer,” *IET Control Theory & Applications*, vol. 13, no. 18, pp. 3096–3106, 2019.
- [7] J. F. Fisac, A. K. Akametalu, M. N. Zeilinger, S. Kaynama, J. Gillula, and C. J. Tomlin, “A general safety framework for learning-based control in uncertain robotic systems,” *IEEE Transactions on Automatic Control*, vol. 64, no. 7, pp. 2737–2752, 2018.
- [8] M. Maiworm, D. Limon, J. M. Manzano, and R. Findeisen, “Stability of gaussian process learning based output feedback model predictive control,” *IFAC-PapersOnLine*, vol. 51, no. 20, pp. 455–461, 2018.
- [9] F. Smarra, A. Jain, T. De Rubeis, D. Ambrosini, A. D’Annunzio, and R. Mangharam, “Data-driven model predictive control using random forests for building energy optimization and climate control,” *Applied energy*, vol. 226, pp. 1252–1272, 2018.
- [10] M. Canale, L. Fagiano, and M. C. Signorile, “Nonlinear model predictive control from data: a set membership approach,” *International Journal of Robust and Nonlinear Control*, vol. 24, no. 1, pp. 123–139, 2014.
- [11] A. Sukharev, “Optimal method of constructing best uniform approximations for functions of a certain class,” *USSR Computational Mathematics and Mathematical Physics*, vol. 18, no. 2, pp. 21–31, 1978.
- [12] M. Milanese and C. Novara, “Set membership identification of nonlinear systems,” *Automatica*, vol. 40, no. 6, pp. 957–975, 2004.
- [13] J.-P. Calliess, “Conservative decision-making and inference in uncertain dynamical systems,” Ph.D. dissertation, University of Oxford, 2014.
- [14] G. Beliakov, “Interpolation of lipschitz functions,” *Journal of computational and applied mathematics*, vol. 196, no. 1, pp. 20–44, 2006.
- [15] J. M. Manzano, D. Limon, D. Muñoz de la Peña, and J. P. Calliess, “Output feedback MPC based on smoothed projected kinky inference,” *IET Control Theory & Applications*, vol. 13, no. 6, pp. 795–805, 2019.
- [16] D. Limon, T. Alamo, D. M. Raimondo, D. Muñoz de la Peña, J. M. Bravo, and E. F. Camacho, “Input-to-state stability: an unifying framework for robust model predictive control,” in *Proceedings of the Int. Workshop on Assessment and Future Directions of Nonlinear Model Predictive Control (NMPC08)*. 5-9 Sept. Pavia, Italy., 2008.
- [17] J. M. Manzano, D. Limon, D. Muñoz de la Peña, and J. P. Calliess, “Robust data-based model predictive control for nonlinear constrained systems,” *IFAC-PapersOnLine*, vol. 51, no. 20, pp. 505–510, 2018.
- [18] I. Leontaritis and S. A. Billings, “Input-output parametric models for non-linear systems part i: deterministic non-linear systems,” *International journal of control*, vol. 41, no. 2, pp. 303–328, 1985.
- [19] A. Levin and K. Narendra, “Identification of nonlinear dynamical systems using neural networks,” in *Neural systems for control*. Elsevier, 1997, pp. 129–160.
- [20] J.-P. Calliess, “Lazily adapted constant kinky inference for nonparametric regression and model-reference adaptive control,” *arXiv preprint arXiv:1701.00178*, 2016.
- [21] Z. B. Zabinsky, R. L. Smith, and B. P. Kristinsdottir, “Optimal estimation of univariate black-box lipschitz functions with upper and lower error bounds,” *Computers & Operations Research*, vol. 30, no. 10, pp. 1539–1553, 2003.
- [22] J. B. Rawlings and D. Q. Mayne, *Model Predictive Control: Theory and Design*, 1st ed. Nob-Hill Publishing, 2009.
- [23] M. Fiacchini, T. Alamo, and E. F. Camacho, “On the computation of convex robust control invariant sets for nonlinear systems,” *Automatica*, vol. 46, no. 8, pp. 1334–1338, 2010.
- [24] G. Calafiore and F. Dabbene, *Probabilistic and randomized methods for design under uncertainty*. Springer, 2006.
- [25] D. Limon, T. Alamo, F. Salas, and E. F. Camacho, “On the stability of MPC without terminal constraint,” *IEEE Transactions on Automatic Control*, vol. 42, pp. 832–836, 2006.
- [26] L. Magni, G. De Nicolao, L. Magnani, and R. Scattolini, “A stabilizing model-based predictive control algorithm for nonlinear systems,” *Automatica*, vol. 37, pp. 1351–1362, 2001.
- [27] L. Ljung, *System identification*. Springer, 1998.